

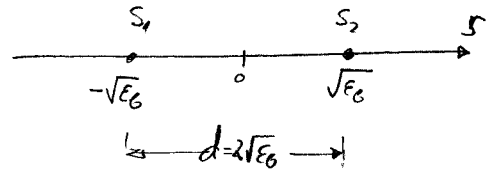
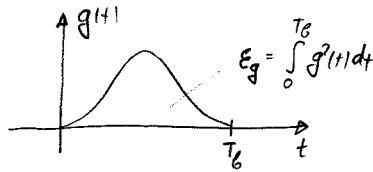
# Performance of The Optimum Receiver for Memoryless Modulation

PAM:  
Binary

$$s_1(t) = g(t)$$

$$s_2(t) = -g(t)$$

↑  
antipodal signalling

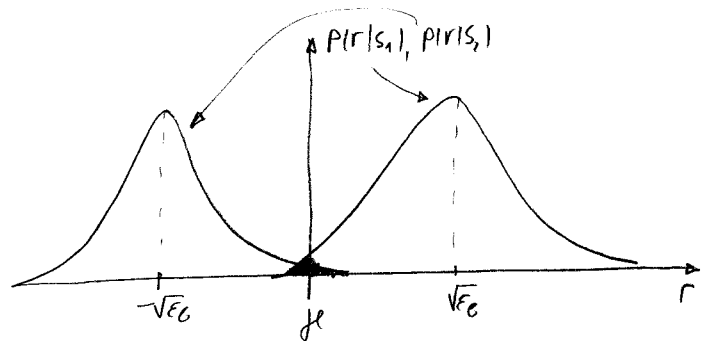


Let us assume that  $P(s_1) = P(s_2)$   
and that  $s_1(t)$  was transmitted

$$r = s_1 + u = \sqrt{E_b} + u \quad \sigma_u^2 = \frac{1}{2} N_0$$

$$p(r|s_1) = \frac{1}{\sqrt{\pi N_0}} e^{-\frac{(r-\sqrt{E_b})^2}{N_0}}$$

$$p(r|s_2) = \frac{1}{\sqrt{\pi N_0}} e^{-\frac{(r-\sqrt{E_b})^2}{N_0}}$$



$$P_e = P(s_1) \cdot P(s_2|s_1) + P(s_2) \cdot P(s_1|s_2)$$

$$P_e = \frac{1}{2} \cdot 2 \cdot P(s_2|s_1) = P(r < \mu) = \int_{-\infty}^{\mu} p(r|s_1) dr$$

↑ optimal  $\mu = 0$

$$P_e = \frac{1}{\sqrt{\pi N_0}} \int_{-\infty}^0 e^{-\frac{(r-\sqrt{E_b})^2}{N_0}} dr$$

$$P_e = \frac{1}{\sqrt{\pi N_0}} \int_{-\infty}^{-\frac{\sqrt{2E_b}}{N_0}} e^{-\frac{x^2}{2}} dx = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{-\frac{\sqrt{2E_b}}{N_0}} e^{-x^2/2} dx$$

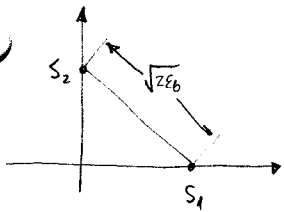
$$P_e = P_b = Q\left(\sqrt{\frac{2E_b}{N_0}}\right)$$

$\frac{E_b}{N_0}$  - signal to noise ratio per bit

$d = 2\sqrt{E_b}$  - distance

$$P_b = Q\left(\sqrt{\frac{d^2}{2N_0}}\right)$$

# Orthogonal signals



$$s_1 = (\sqrt{E_b}, 0)$$

$$s_2 = (0, \sqrt{E_b})$$

$$d = \sqrt{2E_b}$$

Assume that  $s_1$  was transmitted.

$$\bar{r} = (\sqrt{E_b} + n_1, n_2)$$

$$C(\bar{r}, \bar{s}_1) = 2\bar{r}\bar{s}_1 - \|\bar{s}_1\|^2 = 2(\sqrt{E_b} + n_1, n_2) \cdot (\sqrt{E_b}, 0) - E_b = 2E_b + 2\sqrt{E_b}n_1 - E_b = E_b + 2\sqrt{E_b}n_1$$

$$C(\bar{r}, \bar{s}_2) = 2\bar{r}\bar{s}_2 - \|\bar{s}_2\|^2 = 2(\sqrt{E_b} + n_1, n_2) \cdot (0, \sqrt{E_b}) - E_b = 2\sqrt{E_b}n_2 - E_b = -E_b + 2\sqrt{E_b}n_2$$

$$P(e|\bar{s}_1) = P_r \{ C(\bar{r}, \bar{s}_2) > C(\bar{r}, \bar{s}_1) \}$$

$$= P_r \left\{ n_2 - \frac{1}{2}\sqrt{E_b} > n_1 + \frac{1}{2}\sqrt{E_b} \right\}$$

$$= P_r \{ n_2 - n_1 > \sqrt{E_b} \}$$

$$= \frac{1}{\sqrt{2\pi}N_0} \int_{\frac{\sqrt{E_b}}{\sqrt{2\pi}N_0}}^{+\infty} e^{-\frac{x^2}{2N_0}} dx$$

$$= \frac{1}{\sqrt{2\pi}} \int_{\frac{\sqrt{E_b}}{N_0}}^{+\infty} e^{-\frac{x^2}{2}} dx$$

$$= Q\left(\sqrt{\frac{E_b}{N_0}}\right)$$

$$P_B = Q\left(\sqrt{\frac{E_b}{N_0}}\right)$$

$$n_1: N(0, \frac{N_0}{2})$$

$$n_2: N(0, \frac{N_0}{2})$$

$$n_2 - n_1: N(0, N_0)$$

antipodal

$$P_B = Q\left(\sqrt{\frac{E_b}{2N_0}}\right)$$

$$d = 2\sqrt{E_b}$$

orthogonal

$$P_B = Q\left(\sqrt{\frac{E_b}{N_0}}\right)$$

$$d = \sqrt{2E_b}$$

orthogonal signals require increase of energy to achieve the same  $P_B$

$$10 \log_{10} 2 = 3 \text{ dB} \quad - \text{SNR loss is 3 dB}$$

why? - the distance is smaller

# Probability of Error for M-ary Orthogonal Signals

- equal energies, equal probabilities
- $S_m$   $1 \leq m \leq M$

$$C(\bar{r}, \bar{s}_m) = 2\bar{r}\bar{s}_m - \|\bar{s}_m\|^2$$

$$C'(\bar{r}, \bar{s}_m) = 2\bar{r}\bar{s}_m$$

- Assume the  $s_1$  was transmitted

$$\Gamma = (\sqrt{E_s} + n_1, n_2, \dots, n_M) \quad \sigma_n = \frac{1}{2} N_0, \quad n_1, n_2, \dots, n_M \text{ independent, } N(0, \frac{N_0}{2})$$

$$C'(\bar{r}, \bar{s}_1) = (\sqrt{E_s} + n_1, n_2, \dots, n_M) (\sqrt{E_s}, 0, \dots, 0) = E_s + \sqrt{E_s} n_1$$

$$C'(\bar{r}, \bar{s}_2) = (\sqrt{E_s} + n_1, n_2, \dots, n_M) (0, \sqrt{E_s}, \dots, 0) = \sqrt{E_s} n_2$$

$$\vdots$$

$$C'(\bar{r}, \bar{s}_M) = (\sqrt{E_s} + n_1, n_2, \dots, n_M) (0, 0, \dots, \sqrt{E_s}) = \sqrt{E_s} n_M$$

correlator outputs:

$$y_1 = \sqrt{E_s} + n_1$$

$$y_2 = n_2$$

$$\vdots$$

$$y_M = n_M$$

$$P_{y_1}(y_1) = \frac{1}{\sqrt{\pi N_0}} e^{-\frac{(y_1 - \sqrt{E_s})^2}{N_0}}$$

$$P_{y_m}(y_m) = \frac{1}{\sqrt{\pi N_0}} e^{-\frac{y_m^2}{N_0}} \quad 2 \leq m \leq M$$

Probability of correct decision

$$P_c = \int_{-\infty}^{+\infty} P(y_1) P(n_2 < y_1, n_3 < y_1, \dots, n_M < y_1 | y_1) dy_1$$

all other samples in other dimensions should be smaller than  $y_1$

$$P(n_m < y_1 | y_1) = \int_{-\infty}^{y_1} P_{y_m}(x) dx = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{y_1} e^{-\frac{x^2}{N_0}} dx$$

$$P_c = \int_{-\infty}^{+\infty} \prod_{m=2}^M P(n_m < y_1 | y_1) P(y_1) dy_1 = \int_{-\infty}^{+\infty} \left( \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{y_1 \sqrt{2/N_0}} e^{-\frac{x^2}{2}} dx \right)^{M-1} P(y_1) dy_1$$

$P_M = 1 - P_c$  - probability of error

$$P_M = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{+\infty} \left( 1 - \left( \frac{1}{\sqrt{2\pi}} \int_{-\infty}^y e^{-\frac{x^2}{2}} dx \right)^{M-1} \right) e^{-\frac{1}{2} \left( y - \sqrt{\frac{2E_s}{N_0}} \right)^2} dy$$

Symbol error probability

$$E_s = k \cdot E_b \quad 2^k = M$$

energy per symbol      energy per bit

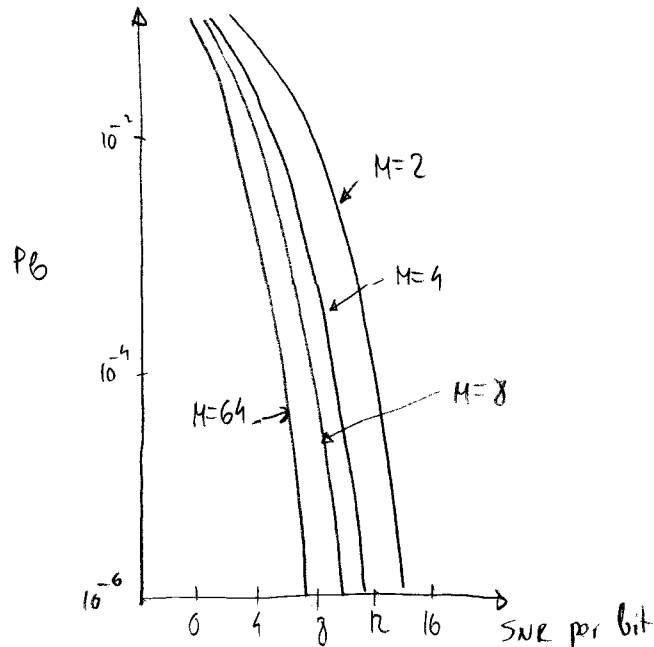
particular Symbol error probability =  $\frac{P_M}{M-1}$

$\binom{K}{j}$  - combinations in which  $j$  bits out of  $K$  may be in error

The average number of bit errors per symbol ( $k$ -bits)

$$\sum_{j=0}^k j \binom{K}{j} \frac{P_M}{M-1} = k \frac{2^{k-1}}{2^k - 1} P_M$$

$$P_B = \frac{2^{k-1}}{2^k - 1} P_M \approx \frac{P_M}{2} \quad k \gg 1$$



# Union Bound on the Probability of Error

What happens when  $M$  increases

$E_i$  - event that  $C(\bar{r}, \bar{s}_i) > C(\bar{r}, \bar{s}_1)$   $i \neq 1$

$$P_M = P\left(\bigcup_{i=1}^{M-1} E_i\right) \leq \sum_{i=1}^{M-1} P(E_i)$$

$$P_M \leq (M-1) P_2 = (M-1) Q\left(\sqrt{\frac{E_s}{N_0}}\right) < M Q\left(\sqrt{\frac{E_s}{N_0}}\right)$$

$$Q\left(\sqrt{\frac{E_s}{N_0}}\right) < e^{-\frac{E_s}{2N_0}}$$

$$P_M < M e^{-\frac{E_s}{2N_0}} = 2^k e^{-\frac{kE_b}{2N_0}}$$

$$< e^{-k(E_b/N_0 - 2 \ln 2)}$$

$$k \rightarrow \infty \quad (M \rightarrow \infty) \quad P_M \rightarrow 0 \quad \text{if} \quad \frac{E_b}{N_0} > 2 \ln 2$$

$$\frac{E_b}{N_0} > 2 \ln 2 = 1.386 \text{ dB}$$

Not very tight bound

$$P_M < 2 e^{-k\left(\sqrt{\frac{E_b}{N_0}} - \sqrt{\ln 2}\right)^2}$$

$$\frac{E_b}{N_0} > \ln 2 = 0.693 = -1.6 \text{ dB}$$

↑  
Shannon limit